• The series interconnection of the ITI systems with impulse responses h_1 and h_2 is the ITI system with impulse response $h = h_1 * h_2$. That is, we have the equivalences shown below.



• The *parallel* interconnection of the ITI systems with impulse responses h_1 and h_2 is a ITI system with the impulse response $h = h_1 + h_2$. That is, we have the equivalence shown below.



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Section 8.3

Properties of LTI Systems

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• A LTI system with impulse response h is memoryless if and only if

h(n) = 0 for all nj = .0

• That is, a LTI system is memoryless if and only if its impulse response h is of the form

where K is a complex constant.

• Consequently, every memoryless ITI system with input X and output Y is characterized by an equation of the form

$$y = x * (K\delta) = Kx$$

)i.e., the system is an ideal amplifier.(

• For a ITI system, the memoryless constraint is extremely restrictive (as every memoryless ITI system is an ideal amplifier.(

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• AITI system with impulse response h is causal if and only if

$$h(n) = 0 \quad \text{for all } n < 0$$

)i.e., *h* is a causal sequence.(

• It is due to the above relationship that we call a sequence X, satisfying

$$x(n) = 0$$
 for all $n < 0$

a causal sequence.

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- The inverse of a ITI system, if such a system exists, is a ITI system.
- Let h and h_{inv} denote the impulse responses of a ITI system and its (ITI) inverse, respectively. Then:

$$h * h_{inv} = \delta.$$

• Consequently, a ITI system with impulse response h is invertible if and only if there exists a sequence h_{nv} such that

$$h * h_{inv} = \delta.$$

• Except in simple cases, the above condition is often quite difficult to test.

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• A ITI system with impulse response h is BIBO stable if and only if

$$\sum_{n = \infty}^{\infty} |h(n \infty) \rangle |($$

)i.e., *h* is *absolutely summable*.(

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• An input x to a system H is said to be an eigensequence of the system H with the eigenvalue λ if the corresponding output y is of the form

$$y = \lambda x$$

where λ is a complex constant.

- In other words, the system H acts as an ideal amplifier for each of its eigensequences x, where the amplifier gain is given by the corresponding eigenvalue λ .
- Different systems have different eigensequences.
- Of particular interest are the eigensequences of (DT) ITI systems.

- As it turns out, every complex exponential is an eigensequence of all ITI systems.
- For a ITI system H with impulse response h,

$$H\{z^{n}\} = H(z)z^{n},$$

where Z is a complex constant and

$$H(Z=(\sum_{n=1}^{\infty}h(n)Z^{-n}.$$

- That is, Z^n is an eigensequence of a LTI system and H(Z) is the corresponding eigenvalue.
- We refer to *H* as the system function (or transfer function) of the system *H*.
- From above, we can see that the response of a ITI system to a complex exponential is the same complex exponential multiplied by the complex factor H(z(

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- Consider a ITI system with input X, output Y, and system function H.
- Suppose that the input X can be expressed as the linear combination of complex exponentials

$$x(n) = \sum_{k} a_{k} z_{k}^{n}$$

where the A_k and Z_k are complex constants.

Using the fact that complex exponentials are eigenfunctions of ITI systems, we can conclude

$$y(n) = \sum_{k} a_{k} H(z_{k}) Z_{k}^{n}$$

- Thus, if an input to a ITI system can be expressed as a linear combination of complex exponentials, the output can also be expressed as linear combination of the *same* complex exponentials.
- The above formula can be used to determine the output of a ITI system from its input in a way that does not require convolution.

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Part 9

Discrete-Time Fourier Series (DTFS(

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- The Fourier series is a representation for *periodic* sequences.
- With a Fourier series, a sequence is represented as a *linear combination of complex sinusoids*.
- The use of complex sinusoids is desirable due to their numerous attractive properties.
- Perhaps, most importantly, complex sinusoids are *eigensequences* of (DT) ITI systems.

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Section 9.1

Fourier Series

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- A set of periodic complex sinusoids is said to be harmonically related if there exists some constant $2\pi / N$ such that the fundamental frequency of each complex sinusoid is an integer multiple of $2\pi / N$.
- Consider the set of harmonically-related complex sinusoids given by

 $\varphi_k(n) = e^{i(2\pi / N)kn}$ for all integer k.

• In the above set $\{\varphi_k\}$, only N elements are distinct, since

 $\varphi_k = \varphi_{k+N}$ for all integer *k*.

• Since the fundamental frequency of each of the harmonically-related complex sinusoids is an integer multiple of $2\pi \frac{1}{N}$ linear combination of these complex sinusoids must be *N*-periodic.

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- A periodic complex-valued sequence X with fundamental period N can be represented as a linear combination of harmonically-related complex sinusoids as
 - •where $\sum_{k=(N)} denotes$ summation over any N consecutive integers (e.g., 0, 1, ..., N-1). (The summation can be taken over any N consecutive integers, due to the N-periodic nature of X and $\Theta^{i(2\pi N)kn}$.) •The above representation of X is known as the (DT) Fourier series
- and the *a_k* are called Fourier series coefficients.
 The above formula for *X* is often called the Fourier series
- synthesis equation.
- The terms in the summation for k = K and k = -K are called the Kth
- harmonic components, and have the fundamental frequency $K(2\pi/N)$. To denote that the sequence X has the Fourier series
- coefficient sequence *a*, we write

$$X(n) \leftarrow \mathcal{A}_{k}$$

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A periodic sequence x with fundamental period N has the Fourier series coefficient sequence a given by

$$a_k = \frac{1}{N} \sum_{n=(N)} x(n) e^{-j(2\pi N) kn}.$$

(The summation can be taken over any *N* consecutive integers due to the *N*-periodic nature of *X* and $e^{-j(2\pi N)kn}$.)

- The above equation for a_k is often referred to as the Fourier series analysis equation.
- Due to the N-periodic nature of X and e^{-j(2π/N)kn}, the sequence a is also N-periodic.

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